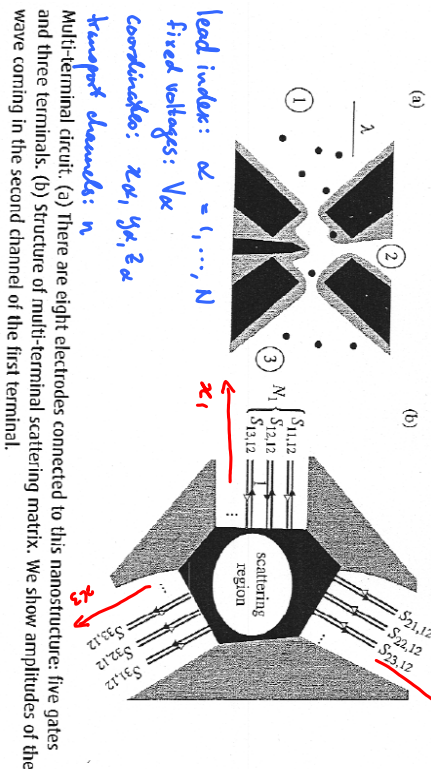
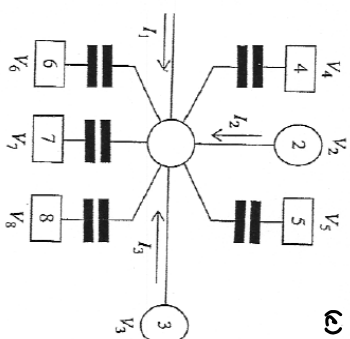


1.5 Multi-terminal circuits (MTC)



lead index: $\alpha = 1, \dots, N$
 fixed voltages: V_α
 coordinates: $x_\alpha, y_\alpha, z_\alpha$
 transport channels: n
 Multi-terminal circuit: (a) There are eight electrodes connected to this nanostructure: five gates and three terminals. (b) Structure of multi-terminal scattering matrix. We show amplitudes of the wave coming in the second channel of the first terminal.



Circuit diagram

$$\psi(x, y, z) = \sum_n \frac{1}{\sqrt{a_n v_n}} \Phi(y_\alpha, z_\alpha) [a_{Rn} e^{-ik_x^{(\alpha n)} x_R} + b_{Rn} e^{+ik_x^{(\alpha n)} x_R}] \quad (1)$$

generate.

$$\psi(x, y, z) = \sum_n \frac{1}{\sqrt{a_n v_n}} \Phi(y_\alpha, z_\alpha) [a_{\alpha n} e^{-ik_x^{(\alpha n)} x_\alpha} + b_{\alpha n} e^{+ik_x^{(\alpha n)} x_\alpha}] \quad (2)$$

incoming

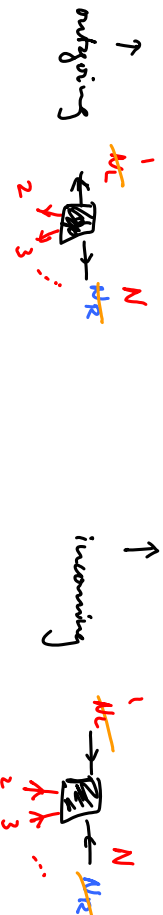
outgoing

$$k_x^{(\alpha n)} = \sqrt{2m(E - E_{\alpha n})/\hbar} \quad (3)$$

(MTC2)

$$b_{\alpha n} = \sum_{\beta=1, \dots, N} \sum_{m=1, \dots, N_\beta} S_{\alpha n, \beta m} a_{\beta m} \quad \alpha = L, R, 1, \dots, N \quad (1)$$

$m = 1, \dots, N_\beta$



Unitarity:

$$\hat{S}^\dagger \hat{S} = \hat{1}, \quad \hat{S} \hat{S}^\dagger = \hat{1} \quad (2)$$

$\hat{S}_{\alpha\alpha}$: reflection in lead α (3)

$\hat{S}_{\alpha\beta}$: transmission from lead $\beta \rightarrow$ lead α (4)

Time reversibility: $\hat{S}(\beta) = \hat{S}^\dagger(-\beta) \xrightarrow{\text{generating}} S_{\alpha n, \beta m}(\beta) = S_{\beta m, \alpha n}(-\beta) \quad (5)$

7.5.2010 (MTC1)

Multi-terminal Landauer formula



(MTC3)

Current in lead α ,
at $x_\alpha = \infty$:

$$I_\alpha = \frac{1}{2e} \sum_n \int_{-\infty}^{\infty} \frac{dk_x}{2\pi} v_x(k_x) f(k_x) f(k_x) \quad (1)$$

come from	with probability	$f_n(k_x)$
$k_x < 0$ $k_x > 0$	α_n	$f_\alpha(E)$
	β_m	$f_\beta(E)$
	$ S_{\alpha n, \beta m} ^2$	

$$I_\alpha = \frac{1}{2e} \sum_n \int_{-\infty}^{\infty} \frac{dk_x}{2\pi} v_x(k_x) [-f_\alpha(E) + \sum_m |S_{\alpha n, \beta m}|^2 f_\beta(E)] \quad (4)$$

$$= -\frac{1}{e} g_\alpha \sum_n \int dE \sum_\beta [\delta_{\alpha\beta} \delta_{nn} - |S_{\alpha n, \beta m}|^2] f_\beta(E) \quad (5)$$

$$I_\alpha = -\frac{1}{e} g_\alpha \int dE \sum_\beta \text{Tr} \{ \delta_{\alpha\beta} - \hat{S}_{\beta\alpha}^\dagger \delta_{\alpha\beta} \} f_\beta(E) \quad (6)$$

← over channels

Current conservation: (total current in all terminals vanishes) (MTC4)

$$\sum_\alpha I_\alpha = -\frac{1}{e} g_\alpha \int dE \sum_\beta \text{Tr} \{ 1 - \underbrace{(\hat{S}^\dagger \hat{S})_{\beta\beta}}_{\text{unitarity: } = 1} \} f_\beta(E) = 0 \quad (1)$$

Linear regime: apply small voltage only to one lead, say γ

$$\mu_\gamma = \epsilon_F + eV_\gamma, \quad \mu_\beta = \epsilon_F \quad \forall \beta \neq \gamma \quad (2)$$

$$f_\beta(E) = [e^{\beta(E - \mu_\beta)} + 1]^{-1} \quad \text{gives 0, since } \text{Tr} (1 - (\hat{S}^\dagger \hat{S})_{\alpha\alpha}) = 0$$

$$I_\alpha = -\frac{1}{e} g_\alpha \int dE \sum_\beta \text{Tr} \{ \delta_{\alpha\beta} - \hat{S}_{\beta\alpha}^\dagger \delta_{\alpha\beta} \} (f_\beta(E) - f(E)) \quad (3)$$

if $\hat{S}_{\alpha\beta}$ is E-independent on scale of eV:

$$I_\alpha = -g_\alpha \text{Tr} [\delta_{\alpha\gamma} - \hat{S}_{\gamma\alpha}^\dagger \hat{S}_{\alpha\gamma}] V_\gamma = g_{\alpha\gamma} V_\gamma \quad \text{"multi-terminal Landauer formula"} \quad (4)$$

In linear regime, contributions to I_α from all leads with $V_\beta \neq 0$ add up: (MTC5)

$$I_\alpha = \sum_\beta g_{\alpha\beta} V_\beta \quad [g_{\alpha\beta}: \text{elements of "conductance matrix"}] \quad (1)$$

Current conservation: $0 = \sum_\alpha I_\alpha$ for arbitrary $V_\beta \Rightarrow \sum_\alpha g_{\alpha\beta} = 0 \quad \forall \beta \quad (2)$

Time reversibility (2.5) implies: $S_{\alpha\mu}, S_{\mu\alpha}(\mathcal{B}) = S_{\beta\mu}, S_{\mu\beta}(-\mathcal{B})$

$g_{\alpha\beta}(\mathcal{B}) = g_{\beta\alpha}(-\mathcal{B})$ "Onsager" relations (3)

For $\mathcal{B} = 0$, $g_{\alpha\beta}$ is symmetric; # of independent parameters then:

(2) $\Rightarrow$ only $(N-1)$ indep. parameters per column.

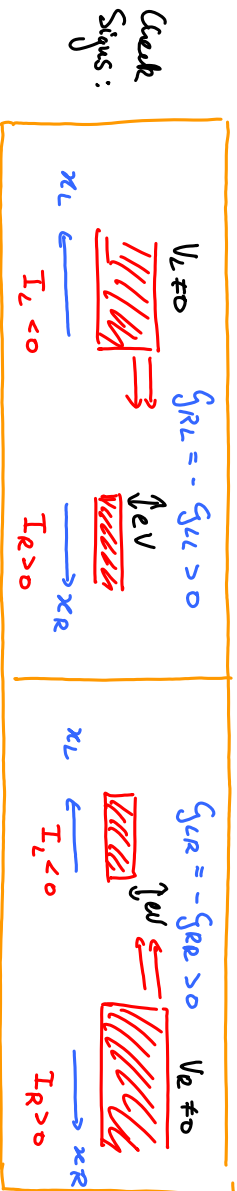
$(N-1)$ diagonal elements, and $[(N-1)^2 - (N-1)]/2$ off-diag. are independent. *since g is symmetric*

In total: $[(N-1)^2 + (N-1)]/2 = N(N-1)/2 \quad (3)$

For $N=2$: only 1 independent element:

(MTC6)

$$g = g_{LR} = g_{RL} = -g_{LL} = -g_{RR}$$



Resistance matrix $R_{\alpha\beta}$:

defined by: $V_\alpha = \sum_\beta R_{\alpha\beta} I_\beta \xRightarrow{(5.1)} R = g^{-1} \quad (1)$

$R_{\alpha\beta}(\mathcal{B}) = R_{\beta\alpha}(-\mathcal{B})$ "Onsager relations" (2)

Voltage probes and two-terminal measurements

(R. Landauer)

MITC7

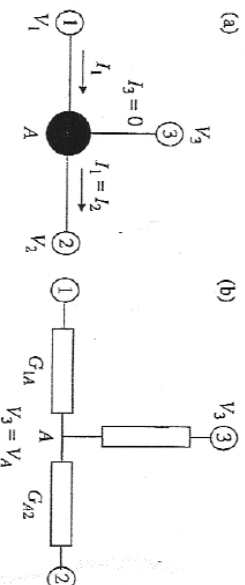
Ideal voltage probe:

non-invasive, does not draw current!

$$I_d^{(5.1)} = \sum_p g_{dp} V_p$$

$$0^{(5.2)} \leq \sum_p g_{dp}$$

(a) Three-terminal circuit with terminal 3 as a voltage probe. (b) A classical circuit could be presented in this way, and conductances of the elements g_{1A} , g_{A2} can be determined from the voltage V_3 measured.



$$0 = I_3 = g_{31} V_1 + g_{32} V_2 + \tilde{g}_{33} V_3 = g_{31}(V_1 - V_3) + g_{32}(V_2 - V_3) \quad (1)$$

for ideal voltmeter

$$\text{Solve: } V_3 = \frac{g_{31} V_1 + g_{32} V_2}{g_{31} + g_{32}} \quad (4.5) = \frac{V_1 T_r(\hat{S}_{12}^T \hat{S}_{31}) + V_2 T_r(\hat{S}_{22}^T \hat{S}_{32})}{T_r(\hat{S}_{12}^T \hat{S}_{31}) + T_r(\hat{S}_{22}^T \hat{S}_{32})} \quad (2)$$

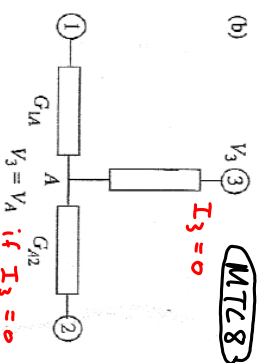
"Noninvasive measurement" requires: $g_{31}, g_{32} \ll g_{12}$, or $\hat{S}_{31}, \hat{S}_{32} \rightarrow 0$.
Then V_3 remains finite! (2)

Finite V_3 is also obtained in classical circuit model:

Consider g_{1A} and g_{A2} in series, $V_3 = V_A$, then

$$\frac{g_{1A}}{g_{12}} = \frac{V_1 - V_2}{V_1 - V_A} \quad (1a) \quad \frac{g_{A2}}{g_{12}} = \frac{V_2 - V_1}{V_2 - V_A} \quad (1b)$$

$\Rightarrow g_{1A}, g_{A2}$ can be measured by measuring V_1, V_2, V_A .



(b)

MITC8

Considering clocks: Solve for V_A :

compare with (3.2)

$$\left. \begin{aligned} (V_1 - V_A) g_{1A} &= g_{12}(V_1 - V_2) \Rightarrow V_A = \frac{1}{g_{1A}} [V_1 (g_{1A} - g_{12}) + g_{12} V_2] \\ (V_2 - V_A) g_{A2} &= g_{A2}(V_2 - V_1) \Rightarrow V_A = \frac{1}{g_{A2}} [V_2 (g_{A2} - g_{12}) + g_{12} V_1] \end{aligned} \right\} \neq V_3 \text{ of (3.2)!} \quad (2a)$$

$$(2a) = (2b) \Rightarrow R_{1A} \left[V_1 \left(\frac{1}{R_{1A}} - \frac{1}{R_{12}} \right) + \frac{V_2}{R_{12}} \right] = R_{A2} \left[V_2 \left(\frac{1}{R_{A2}} - \frac{1}{R_{12}} \right) + \frac{V_1}{R_{12}} \right] \quad (3)$$

$$\Rightarrow V_1 \left[1 - \frac{R_{1A} + R_{A2}}{R_{12}} \right] = V_2 \left[1 - \frac{R_{A2} + R_{1A}}{R_{12}} \right] \quad (4)$$

$$(4) \text{ must hold for any } V_1, V_2 \Rightarrow R_{1A} + R_{A2} = R_{12} \quad \checkmark$$

Difference in descriptions:

MTG 9

Classical: we can apply a voltage to any point in circuit, including A

Quantum: we can apply voltage only to reservoir; if A represents a nanostructure (e.g. single-channel conductor), it is too small to apply a voltage directly to it.

Lets see what happens if we try to assign voltages "inside a nanostructure":

If we write $R_{\text{tot}} = R_{1A} + R_{AB} + R_{B2}$, what is R_{AB} ?

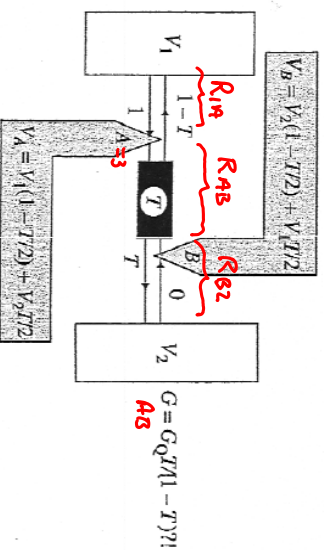
In this classical picture, we would have (in analogy to (3.1)):

$$g_{1A} = g_{12} \frac{V_1 - V_2}{V_1 - V_A} \quad (1)$$

$$g_{AB} = g_{12} \frac{V_1 - V_2}{V_A - V_B} \quad (2)$$

$$g_{B2} = g_{12} \frac{V_1 - V_2}{V_B - V_2} \quad (3)$$

So: First determine V_A, V_B , then we get g_{AB} from (2).



Let us get $V_A = V_3$ from (3.2):

$$V_A = V_3 \stackrel{(3.2)}{=} \frac{g_{31} V_1 + g_{32} V_2}{g_{31} + g_{32}} \quad (1)$$

We first need:

$$g_{31} = g_{A1} = g_{A2} [\omega + (1-T)(1-\omega)\omega] \approx g_{A2} \omega (2-T) \quad (2a)$$

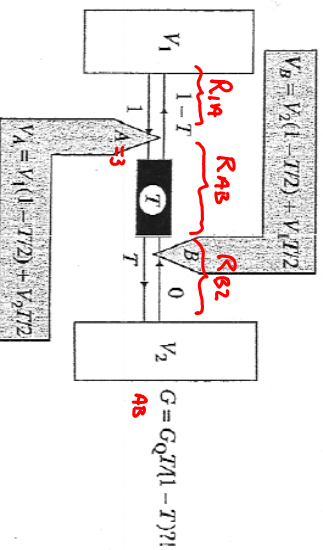
$$g_{32} = g_{A2} = g_{A2} \omega T \quad (2b)$$

$$(2) \text{ into (1): } V_A = \frac{(2-T)V_1 + T V_2}{(2-T) + T} = (1-T/2)V_1 + T/2 V_2 \neq V_1 \quad (3a)$$

$$\text{Similarly: } V_B = (1-T/2)V_2 + T/2 V_1 \neq V_2 \quad (3b)$$

That $V_A \neq V_1$ is a consequence of pretending that voltage V_A can be applied directly inside nanostructure ...

MTG 9
25: probability to enter A



Let us proceed nevertheless and insert V_A, V_B from (2.2) into (4.3): MTC 4

$$1 \rightarrow A: \quad g_{1A} \stackrel{(4.1)}{=} g_{12} \frac{V_1 - V_2}{V_1 - V_A} \stackrel{(3.2a)}{=} g_0 T \frac{V_1 - V_2}{V_1 - [x - \pi/2)V_1 + \pi/2 V_2]} = 2g_0 \quad (1)$$

$$A \rightarrow B: \quad g_{AB} \stackrel{(4.2)}{=} g_{12} \frac{V_1 - V_2}{V_A - V_B} \stackrel{(3.3)}{=} g_0 T \frac{V_1 - V_2}{[(1 - \pi/2)V_1 + \pi/2 V_2] - [x - \pi/2)V_1 + \pi/2 V_2]} \\ = \frac{g_0 T}{1 - T} \quad (\neq g_0 T = \text{two-terminal Landauer formula!}) \quad (2)$$

$$B \rightarrow 2: \quad g_{B2} \stackrel{(4.3)}{=} g_{12} \frac{V_1 - V_2}{V_B - V_2} = g_0 T \frac{V_1 - V_2}{[x - \pi/2)V_1 + \pi/2 V_2] - V_2} = 2g_0 \quad (3)$$

Consistency check:

$$R_{\text{tot}} = R_{1A} + R_{AB} + R_{B2} = \frac{1}{g_{1A}} + \frac{1}{g_{AB}} + \frac{1}{g_{B2}} = \frac{1}{g_0} \left[\frac{1}{2} + \frac{1-T}{T} + \frac{1}{2} \right] = \frac{1}{g_0 T} = \frac{1}{g_{\text{tot}}} \quad (4)$$

Comments:

$$g_{AB} = \frac{g_0 T}{1-T} \quad \neq g_0 T \quad \text{as expected from Landauer formula} \quad \text{MTC 15}$$

= "wrong Landauer formula"

$$\bullet \text{ For QPC with } T=1: \quad g_{AB} \rightarrow \infty \quad \text{or} \quad R_{AB} = g_{AB}^{-1} \rightarrow 0$$

This does make sense: QPC has no resistance and no voltage drop across it.

$$\bullet \text{ Voltage drops across } R_{1A} = g_{1A}^{-1} = \frac{1}{2g_0} \quad \text{and} \quad R_{1B} = g_{1B}^{-1} = \frac{1}{2g_0} \\ \text{(called "contact resistances" by Landauer),}$$

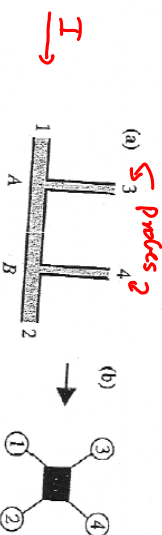
• What went wrong: in reality, it is not possible to apply voltages V_A and V_B to a single-channel scatterer. Voltages can be applied only to reservoirs. It is not possible to separate nanostructure into separate elements of definite resistance.

$$\bullet \text{ Tunnel junction: } T \ll 1: \quad g_{AB} \approx g_0 T = g_{\text{tot}} \quad \checkmark \quad \text{as expected.}$$

Reason: voltage drops across tunnel junction, contribution of contact resistances is ≈ 0

If Eqs. (11.1) - (11.3) are to hold, we need a 4-terminal circuit:

(MTC13)



A common scheme for two-probe measurement of the resistance R_{AB} . (a) does not generally work in quantum transport. The reason is that the equivalent four-terminal nanostucture (b) cannot be generally separated into elements of definite resistance.

Classically: terminals 1 & 2 pass current

terminals 3 & 4 probe voltage drop between A & B

$$\text{we expect } R_{AB} = \frac{V_3 - V_4}{I}$$

Quantum transport: this does not work in general!

Eg: assume non-inverse voltage probes:

$$g_{43}, g_{34} \ll g_{41}, g_{42} \\ g_{31}, g_{32} \ll g_{12}, g_{21}$$

$$\text{Using } I_\alpha = \sum_{\beta} g_{\alpha\beta} V_{\beta}, \quad 0 = \sum_{\beta} g_{\alpha\beta} V_{\beta}, \quad \text{(MTC14)} \quad (1)$$

we get from

$$0 = I_3 = g_{31}V_1 + g_{32}V_2 + g_{33}V_3$$

$$0 = I_4 = g_{41}V_1 + g_{42}V_2 + g_{44}V_4$$

$$V_3 - V_4 = (V_1 - V_2) \left(\frac{g_{31}}{g_{31} + g_{32}} - \frac{g_{41}}{g_{41} + g_{42}} \right) \quad (2)$$

$$\text{Current: } I = g_{21}(V_1 - V_2) \quad (3)$$

$$R = \frac{V_3 - V_4}{I} = \frac{1}{g_{21}} \left(\frac{g_{31}}{g_{31} + g_{32}} - \frac{g_{41}}{g_{41} + g_{42}} \right) \neq R_{AB} \quad (4)$$

Note also $R(8) \neq R(-8)$: Onsager relations do not hold in multiterminal setup.